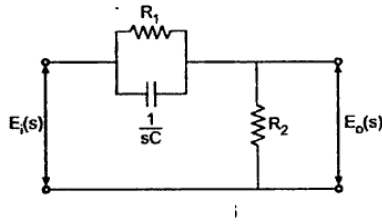


Ans Q#1:

Solution : Take Laplace transform of the network, as shown in the Fig



The parallel combination of R_1 and $\frac{1}{sC}$ gives impedance of,

$$Z = \frac{Z_1 Z_2}{Z_1 + Z_2} = \frac{R_1 \times \frac{1}{sC}}{R_1 + \frac{1}{sC}} = \frac{R_1}{1 + s R_1 C}$$

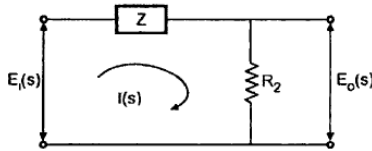


Fig. 3.17

Applying KVL to the circuit,

$$E_i(s) = Z I(s) + I(s) R_2 \quad \dots (1)$$

$$E_o(s) = I(s) R_2 \quad \dots (2)$$

$$\therefore I(s) = \frac{E_o(s)}{R_2} \quad \text{from (2)}$$

Substituting in (1) we get,

$$E_i(s) = I(s) [Z + R_2] = \frac{E_o(s)}{R_2} [Z + R_2]$$

$$\therefore \frac{E_o(s)}{E_i(s)} = \frac{R_2}{Z + R_2}$$

$$\text{Substituting } Z, \text{ T. F.} = \frac{R_2}{\frac{R_1}{1 + s R_1 C} + R_2} = \frac{R_2 (1 + s R_1 C)}{R_1 + R_2 (1 + s R_1 C)}$$

$$= \frac{s R_1 R_2 C + R_2}{R_1 + s R_1 R_2 C + R_2} = \frac{s + \alpha}{s + \beta}$$

$$\text{where } \alpha = \frac{1}{R_1 C}, \quad \beta = \frac{(R_1 + R_2)}{R_1 R_2 C}$$

This circuit is also called **lead compensator**.

Ans Q#2:

Solution : Two forward paths, $K = 2$,

$$T_1 = G_1 G_3 G_4 G_5 G_6$$

$$T_2 = G_1 G_2 G_6$$

Loops are, $L_1 = -G_4 H_1$

$$L_2 = -G_3 G_4 G_5 H_2$$

$$L_3 = -G_2 H_2$$

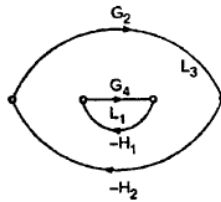


Fig. 6.14(a) Non touching loops

Out of these, L_1 and L_3 is combination of 2 non touching loops

$$\Delta = 1 - [L_1 + L_2 + L_3] + [L_1 L_3]$$

$\Delta_1 =$ Eliminate L_1, L_2, L_3 as all are touching to T_1 from Δ

$\therefore \Delta_1 = 1$

$\Delta_2 =$ Eliminate L_2 and L_3 , as they are touching to T_2 , from

Δ . But L_1 is non touching hence keep it as it is in Δ .

$\therefore \Delta_2 = 1 - [L_1]$

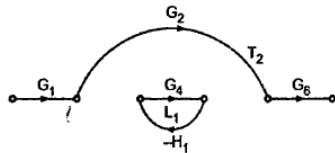


Fig. 6.14(b) L_1 Non touching to T_2

Substitute in Mason's gain formula,

$$T.F. = \frac{T_1 \Delta_1 + T_2 \Delta_2}{\Delta}$$

$$T.F. = \frac{G_1 G_3 G_4 G_5 G_6 [1] + G_1 G_2 G_6 [1 + G_4 H_1]}{1 + G_4 H_1 + G_3 G_4 G_5 H_2 + G_2 H_2 + G_2 G_4 H_1 H_2}$$

Ans Q#3:

Solution : Arrange the given transfer function as,

$$\frac{Q(s)}{I(s)} = \frac{1}{J \left[s^2 + \frac{f}{J}s + \frac{K}{J} \right]} = \frac{\left(\frac{1}{J} \right)}{s^2 + \frac{f}{J}s + \frac{K}{J}}$$

Comparing denominator with $s^2 + 2\xi\omega_n s + \omega_n^2$,

$$\omega_n^2 = \frac{K}{J} \quad \text{i.e.} \quad \omega_n = \sqrt{\frac{K}{J}} \quad \dots (1)$$

$$\text{and} \quad 2\xi\omega_n = \frac{f}{J} \quad \text{i.e.} \quad \xi = \frac{f}{2\sqrt{KJ}} \quad \dots (2)$$

$$\text{Now} \quad M_p = 6\% \quad \text{i.e.,} \quad 0.06$$

$$\therefore 0.06 = e^{-\pi\xi/\sqrt{1-\xi^2}}$$

$$\therefore \ln(0.06) = \frac{-\pi\xi}{\sqrt{1-\xi^2}}$$

$$\text{Solving for } \xi, \quad \xi = 0.667 \quad \dots (3)$$

$$T_p = \frac{\pi}{\omega_d} = \frac{\pi}{\omega_n \sqrt{1-\xi^2}} = 1 \text{ sec}$$

$$\therefore \omega_n = \frac{\pi}{\sqrt{1-(0.667)^2}} = 4.2165 \text{ rad/sec} \quad \dots (4)$$

The Laplace transform of output is Q(s).

Now input is step of 10 Nm hence corresponding Laplace transform is,

$$I(s) = \frac{10}{s}$$

$$\therefore \frac{Q(s)}{\left(\frac{10}{s} \right)} = \frac{1}{Js^2 + fs + K}$$

$$\therefore Q(s) = \frac{10}{s(Js^2 + fs + K)}$$

The steady state of output can be obtained by final value theorem.

$$\text{Steady state output} = \lim_{s \rightarrow 0} sQ(s)$$

$$\therefore 0.5 = \lim_{s \rightarrow 0} \frac{s \cdot 10}{s(Js^2 + fs + K)} = \frac{10}{K}$$

$$\therefore \quad \mathbf{K = 20}$$

$$\text{Equating (1) and (4), } 4.2165 = \sqrt{\frac{K}{J}}$$

$$\therefore \quad 4.2165 = \sqrt{\frac{20}{J}}$$

$$\therefore \quad \mathbf{J = 1.1249}$$

$$\text{From equation (2), } 0.667 = \frac{f}{2\sqrt{KJ}}$$

$$\therefore \quad 0.667 = \frac{f}{2\sqrt{20 \times 1.1249}}$$

$$\therefore \quad \mathbf{f = 6.3274}$$